

DFT Discrete F.T.

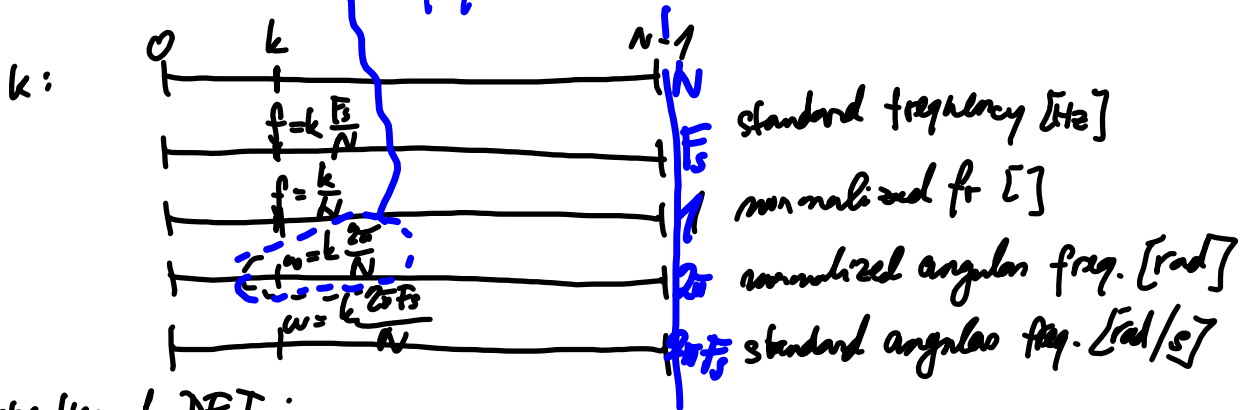
N samples of signal $x[n]$ \rightarrow N samples of spectrum $X[k]$

DFT "analysis" $\left\{ \begin{array}{l} \text{output} = \text{maybe} \\ \text{constant} \end{array} \right.$ $\left\{ \begin{array}{l} \text{sam. input} \\ \text{cp.} \end{array} \right.$ $e^{+j \text{freq. time}}$ IDFT "synthesis"

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi}{N} kn}$$

fundamental freq.

$$x[m] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{+j \frac{2\pi}{N} km}$$



Properties of DFT:

~~$X[k] = X^*[-k]$~~ $X[k] = X^*[N-k]$

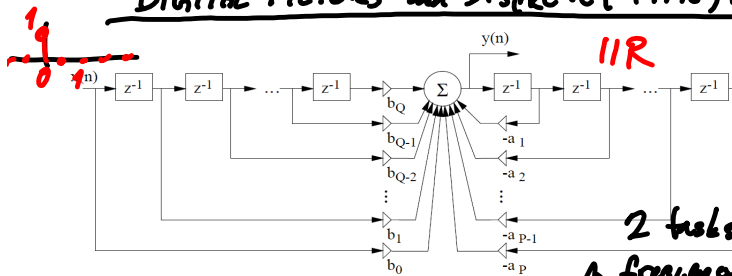
circular shift of signal

$y[m] = x[\text{mod}_N(m-m)]$ $Y[k] = X[k] e^{-j \frac{2\pi}{N} km}$

convolution of 2 signals

$y[m] = x_1[m] \otimes x_2[m]$ $Y[k] = X_1[k] X_2[k]$
 FFT = DFT for $N=2^6$, usually 128, 256, 512, 1024, 2048

DIGITAL FILTERS and DISCRETE-TIME SYSTEMS ...



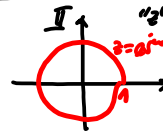
n	0	1	2	3	...	$Q-1$	0	1	2	...
$x[n]$	b_0	b_1	b_2	b_3	...	b_{Q-1}	0	0	...	

- 2 tasks:
- 1) frequency response.
 - 2) stability

THEORY
 Difference equation
 $y[n] = b_0 x[n] + b_1 x[n-1] + \dots + b_Q x[n-Q] - a_1 y[n-1] - \dots - a_P y[n-P]$
 $Y(z) = \sum_{k=0}^Q b_k X(z) z^{-k} - \sum_{k=1}^P a_k Y(z) z^{-k}$

EXAMPLE
 $y[n] = x[n] - 0.5 y[n-1]$
 FT (analog) $X(\omega)$
 $X(j\omega) \approx$ Laplace $X(s)$

Z-transform $X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$



$Y(z) = X(z) - 0.5 Y(z) z^{-1}$
 $Y(z) [1 + 0.5 z^{-1}] = X(z)$
 $\frac{Y(z)}{X(z)} = H(z) = \frac{1}{1 + 0.5 z^{-1}}$

DFT $X(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{j\omega n}$

$x[n] \rightarrow X(z)$ $x[n-1] \rightarrow X(z) z^{-1}$
 $Y(z) = b_0 X(z) + b_1 X(z) z^{-1} + \dots + b_Q X(z) z^{-Q} - a_1 Y(z) z^{-1} - \dots - a_P Y(z) z^{-P}$

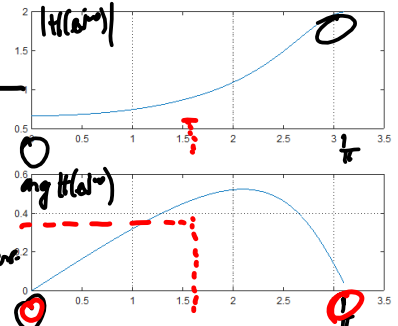
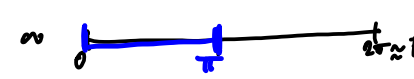
TRANSFER FUNCTION: $H(z) = \frac{Y(z)}{X(z)}$

$H(e^{j\omega}) = \frac{1}{1 + 0.5 e^{j\omega}}$

$Y(z) [1 + a_1 z^{-1} + \dots + a_P z^{-P}] = X(z) [b_0 + b_1 z^{-1} + \dots + b_Q z^{-Q}]$
 $\frac{Y(z)}{X(z)} = \frac{b_0 + b_1 z^{-1} + \dots + b_Q z^{-Q}}{1 + a_1 z^{-1} + \dots + a_P z^{-P}} = H(z) = \frac{\sum_{k=0}^Q b_k z^{-k}}{1 + \sum_{k=1}^P a_k z^{-k}}$

Frequency response
 $H(e^{j\omega}) = H(z) \Big|_{z=e^{j\omega}} = \frac{b_0 + b_1 e^{j\omega} + \dots + b_Q e^{j\omega Q}}{1 + a_1 e^{j\omega} + \dots + a_P e^{j\omega P}} = \frac{\sum_{k=0}^Q b_k e^{j\omega k}}{1 + \sum_{k=1}^P a_k e^{j\omega k}}$

radius
 freq (BA)
 Polhu
 mp freq



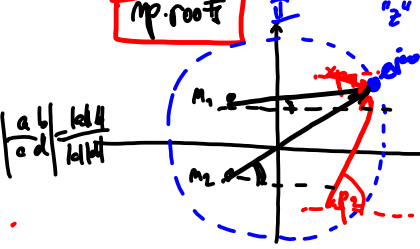
Factorization of transfer function

$H(z) = \frac{b_0 z^Q (z^{-Q} + \frac{b_1}{b_0} z^{-Q+1} + \dots + \frac{b_Q}{b_0})}{z^P (z^{-P} + a_1 z^{-P+1} + \dots + a_P)}$

$ax^2 + bx + c = 0$ $x_{1/2} = 2$ real or 2 complex-conj. numbers

$a(x-x_1)(x-x_2)$
 $H(z) = \frac{b_0 z^Q (z - m_1)(z - m_2) \dots (z - m_M)}{z^P (z - p_1)(z - p_2) \dots (z - p_P)}$

M - roots of numerator "zeros"
 P - roots of the denominator "poles"



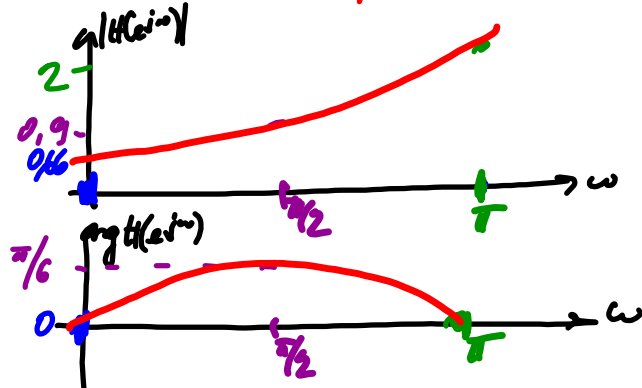
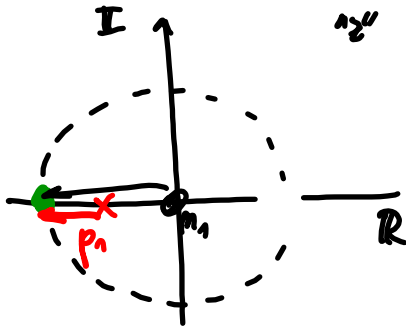
$H(e^{j\omega}) = \frac{b_0 e^{j\omega Q} (e^{j\omega} - m_1)(e^{j\omega} - m_2) \dots}{z^P (e^{j\omega} - p_1)(e^{j\omega} - p_2) \dots}$

$|H(e^{j\omega})| = b_0$ product of lengths of black vectors
 product of lengths of red vectors.

$\arg H(e^{j\omega}) = \omega(P-Q) +$ sum of angles of black vectors
 - sum of angles of red vectors.

$$H(z) = \frac{1}{1+0,5z^{-1}} = \frac{1}{z^1(z+0,5)} = \frac{z-0}{z-(-0,5)}$$

$$H(e^{j\omega}) = \frac{e^{j\omega} - 0}{e^{j\omega} - (-0,5)}$$



$$\begin{aligned} |H| &= \frac{1}{1,5} = 0,66 & \arg H &= 0 - 0 = 0 \\ |H| &= \frac{1}{1,11} = 0,9 & \arg H &= \frac{\pi}{2} - \frac{\pi}{3} = \frac{\pi}{6} \\ |H| &= \frac{1}{0,5} = 2 & \arg H &= \pi - \pi = 0 \end{aligned}$$

